

ON THE OBSERVABILITY AND CONTROLLABILITY CONDITION OF A TEMPERATURE MEASURING SYSTEM

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Abstract

In this paper, we discuss that the necessary and sufficient condition for observability and controllability of a temperature measuring system connected with a proportional approximately derivative action type compensator to improve the dynamic error, where it is assumed that the characteristic of detecting element is delayed in first order.

As a result, it was found that the observability condition is $\alpha_c \neq \alpha_d$, $\kappa_c < \infty$, $\alpha_d \neq 0$, $\alpha_o \neq 0$, $C_1 \neq 0$, $C_2 \neq 0$, and the controllability condition is $\alpha_c \neq \alpha_o$, $\kappa_d \neq 0$, $\kappa_c < \infty$, $\alpha_c \neq 0$, $\alpha_o \neq 0$, for two outputs observing system.

1. Introduction

The transfer function of a detecting part in the temperature measuring system is often approximated at the first order delay system. Hence, if the corner frequency of a detector is lower than the frequency which the true temperature varies, a dynamic error occurs in this system.

For the method which compensates the delay of the detector, the proportional plus incomplete derivative action (PD for short) has been used. For the study of a compensator using PD action, in 1950 Ziegler and Nichols accomplished compensating the time lag of a gas enclosure type temperature detector by a pneumatic PD action¹⁾, and Sawaragi, Wada et al. using a quartz bar and a steel bar for the temperature detector, gave the PD compensation pneumatically by the differential of their expansion²⁾. For electric PD compensators, types using a CR circuit plus a chopper-amplifier³⁾ and a DC amplifier⁴⁾⁵⁾ have been developed.

To discuss the observability and controllability⁶⁾ on the basis of the modern control theory for these PD compensators is significant, although cases which do not necessarily satisfy the observability or controllability condition exist⁷⁾. So we discuss the necessary and sufficient conditions for which the above PD compensators hold good observability and controllability.

2. Notations

E_i [°C] : true temperature (input to the temperature measurement system)

E_d [V] : measured temperature

E_o [V] : output of the compensator (compensation output of temperature measurement system)

I [A] : output of the resistance to current converter (R_t/I)

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K_d [mA/°C] : gain constant of the detector

K_c [mA/V] : gain constant of the compensator

T_d [s] : time constant of the detecting element

T_c [s] : main time constant of the compensator

T_o [s] : subordinate time constant of the compensator (time constant of the incomplete derivation)

$G_d(S)$ [mA/°C] : transfer function of the detecting element

$G_c(S)$ [V/mA] : transfer function of the compensating element

$\kappa_d = K_d / T_d$: gain coefficient of the detecting element

$\kappa_c = T_o K_c / (T_c + T_o)$: gain coefficient of the PD compensator

$\alpha_d = 1 / T_d$: coefficient of the time constant of the detecting element

$\alpha_c = 1 / (T_c + T_o)$: coefficient of the main time constant of the compensator

$\alpha_0 = 1 / T_o$: coefficient of the subordinate time constant of the compensator

C_1, C_2 : conversion coefficient of the observation instrument

3. Compensating method for the temperature detecting lag by PD action

For the compensating system of the detector in Fig. 1, if the transfer function of the detecting element

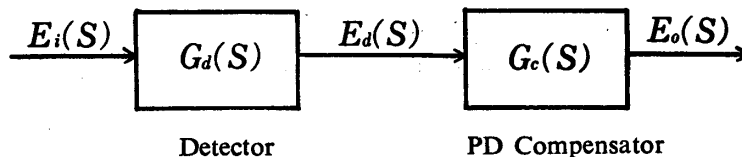


Fig. 1 Block diagram of PD compensating system for temperature detector

$G_d(S)$ is approximated by a first order delayed system and the transfer function of the compensating element $G_c(S)$ is proportional plus imcompleted derivative action, the following equations are obtained

$$G_d(S) = \frac{K_d}{1 + T_d S} \quad (1)$$

$$G_c(S) = \frac{1}{K_c} \left(1 + \frac{T_c S}{1 + T_o S} \right) \quad (2)$$

Now, when to this system the step input voltage $E_i u(t)$ (with $u(t) = 1, t > 0$; $u(t) = 0, t < 0$) is applied, if the conditions $T_o \ll T_d, T_c = T_d, K_c = K_d$ exist, the output voltage $E_o(t)$ is given by the following equation

$$E_o(t) \approx E_i \{ 1 - \exp(-t/T_o) \} \quad (3)$$

If this system has no compensator, the output voltage $E_o(t)$ for $K_d = 1$ is

$$E_o(t) = E_i \{ 1 - \exp(-t/T_d) \} \quad (4)$$

Then the time constant T_d of the detecting element is improved to T_o by the compensating element $G_c(S)$.

4. Design and conditions of the compensator under consideration of observability and controllability

In this section we will discuss the system at general terms. When the scaler input corresponding to the true temperature is $u(t)$ and the state variable of the system $x(t)$ is a second order vector, then b becomes the 2nd order in the state equation $dx(t)/dt = Ax(t) + bu(t)$, and the system in Fig. 2 is organized.

From Fig. 2, the following state equations and observation equations are obtained

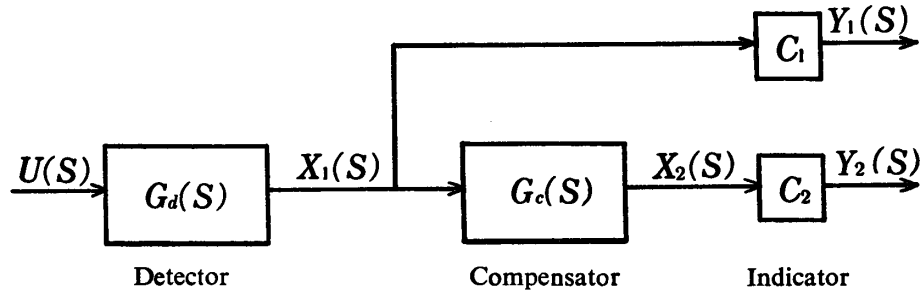


Fig.2 Block diagram of a compensating system under consideration of observability and controllability

$$\left. \begin{aligned} X_1(S) &= G_d(S) \cdot U(S) \\ X_2(S) &= G_c(S) \cdot X_1(S) \end{aligned} \right\} \quad (5)$$

$$Y_1(S) = c_1 X_1(S), \quad Y_2(S) = c_2 X_2(S) \quad (6)$$

and with respect to $G_d(S)$, $G_c(S)$

$$G_d(S) = \frac{K_d}{1 + T_d S} = \frac{\kappa_d}{S + \alpha_d} \quad (7)$$

$$\begin{aligned} G_c(S) &= \frac{1}{K_c} \left(1 + \frac{T_c S}{1 + T_o S} \right) \\ &= \frac{1}{K_c} \frac{(T_o + T_c) S + 1}{1 + T_o S} \\ &= \frac{1}{\kappa_c} \frac{S + \alpha_c}{S + \alpha_o} \end{aligned} \quad (8)$$

where

$$\left. \begin{aligned} \frac{1}{\kappa_c} &= \frac{T_o + T_c}{T_o K_c}, & \kappa_d &= \frac{K_d}{T_d} \\ \alpha_c &= \frac{1}{T_o + T_c}, & \alpha_o &= \frac{1}{T_o} \\ \alpha_d &= \frac{1}{T_d}, & T_c &\gg T_o \end{aligned} \right\} \quad (9)$$

Substituting the equations (7), (8) into the equations (5), (6), the following equations are obtained.

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \end{bmatrix} = \begin{bmatrix} -\alpha_d & 0 \\ (\alpha_c - \alpha_d)/\kappa_c & -\alpha_o \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} \kappa_c \\ \kappa_d/\kappa_c \end{bmatrix} u \quad (10)$$

$$\begin{bmatrix} y_1 \\ y_2 \end{bmatrix} = \begin{bmatrix} c_1 & 0 \\ 0 & c_2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} \quad (11)$$

Therefore, the state equation and the observation equation are

$$\left. \begin{aligned} \frac{dx(t)}{dt} &= Ax(t) + bu(t) \\ y(t) &= Cx(t) \end{aligned} \right\} \quad (12)$$

where

$$\left. \begin{aligned} x(t) &= \text{col} (x_1(t), x_2(t)) \\ y(t) &= \text{col} (y_1(t), y_2(t)) \end{aligned} \right\} \quad (13)$$

$$\left. \begin{aligned} A &= \begin{bmatrix} \alpha_d & 0 \\ (\alpha_c - \alpha_d)/\kappa_c & -\alpha_o \end{bmatrix} \\ b &= \begin{bmatrix} \kappa_c \\ \kappa_d/\kappa_c \end{bmatrix} \\ C &= \begin{bmatrix} c_1 & 0 \\ 0 & c_2 \end{bmatrix} \end{aligned} \right\} \quad (14)$$

Fig. 3 is obtained from the equations (7), (8).

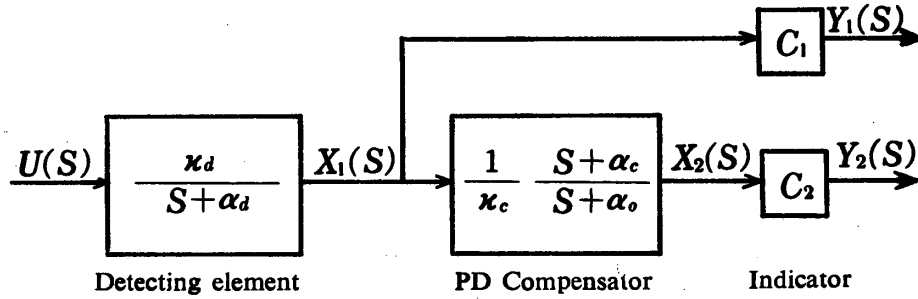


Fig.3 Block diagram of a detecting element and its PD compensating system

The necessary and sufficient condition for observability of the system in Fig. 3 is

$$\text{rank} [C' : A'C'] = 2 \quad (15)$$

where ' denotes the transposition. From equation (14),

$$C' = \begin{bmatrix} c_1 & 0 \\ 0 & c_2 \end{bmatrix} \quad (16)$$

$$\begin{aligned} A'C' &= \begin{bmatrix} -\alpha_d & (\alpha_c - \alpha_d) / \kappa_c \\ 0 & -\alpha_o \end{bmatrix} \begin{bmatrix} c_1 & 0 \\ 0 & c_2 \end{bmatrix} \\ &= \begin{bmatrix} -\alpha_d c_1 & c_2 (\alpha_c - \alpha_d) / \kappa_c \\ 0 & -\alpha_o c_2 \end{bmatrix} \end{aligned} \quad (17)$$

Consequently, the condition which satisfy the following equation

$$\text{rank} \begin{bmatrix} c_1 & 0 & -\alpha_d c_1 & (\alpha_c - \alpha_d) c_2 / \kappa_c \\ 0 & c_2 & 0 & -\alpha_o c_2 \end{bmatrix} = 2, \quad (18)$$

$$\text{are } \alpha_c \neq \alpha_d, \kappa_c < \infty, \alpha_d \neq 0, \alpha_o \neq 0, c_1 \neq 0, c_2 \neq 0 \quad (19)$$

Next, the necessary and sufficient condition for controllability of the system in Fig. 3 is

$$\text{rank} [b : Ab] = 2 \quad (20)$$

From equation (14),

$$\begin{aligned} Ab &= \begin{bmatrix} -\alpha_d & 0 \\ (\alpha_c - \alpha_d) / \kappa_c & -\alpha_o \end{bmatrix} \begin{bmatrix} \kappa_d \\ \kappa_d / \kappa_c \end{bmatrix} \\ &= \begin{bmatrix} -\alpha_d \kappa_d \\ (\alpha_c - \alpha_o - \alpha_d) \kappa_d / \kappa_c \end{bmatrix} \end{aligned} \quad (21)$$

Therefore, the conditions which satisfy the following equation

$$\text{rank} \begin{bmatrix} \kappa_d & -\alpha_d \kappa_d \\ \kappa_d / \kappa_c & (\alpha_c - \alpha_o - \alpha_d) \kappa_d / \kappa_c \end{bmatrix} = 2, \quad (22)$$

$$\text{are } \alpha_c \neq \alpha_o, \kappa_d \neq 0, \kappa_c < \infty, \alpha_c \neq 0, \alpha_o \neq 0 \quad (23)$$

5. Conclusion

- (1) For a PD compensating system which has been organized as a one input and one output system, the system organization shown in Fig. 3 satisfies observability and controllability mathematically and physically.
- (2) We can process data by means of modern analytical methods.
- (3) The values of the coefficients for the electric PD compensator are

$$\begin{aligned} \alpha_d &= 0.05 \text{ s}^{-1}, & \kappa_d &= 0.053 \text{ V}/^\circ\text{C} \\ c_1 &= 18.75 \text{ }^\circ\text{C}/\text{V}, & \alpha_o &= 0.0435 \text{ s}^{-1} \end{aligned}$$

$$\kappa_c = 0.053 \text{ }^\circ\text{C/V}, \quad \alpha_o = 0.3333 \text{ s}^{-1}$$

$$c_2 = 18.75 \text{ }^\circ\text{C/V}$$

These values satisfy all conditions of the equations (19), (20).

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